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of one of the lines (220) forms a further connection 5 for symmetrical signals. The two interconnected ends of the two line couplers form the connecting port 3 for unsymmetrical signals. To achieve a tight coupling (e.g. 3dB) between the lines, the balun is made using multilayer technology. Hence the manufacturing costs are high, which is a problem for the mass-production of, for example, GaAs monolithic integrated circuits.

The balun just described is a single balun and therefore cannot be used in star mixers and modulators and single-sideband mixers. For this purpose it is necessary to use a dual balun. One such device is the subject of a co-pending German patent application filed on 23 February 2000 by Robert Bosch GmbH and is as shown in Figure 2.

The balun of Figure 2 consists of two three-line interdigitated sections. Each section is a quarter-wavelength long. The first and third lines 20, 21 of the first section are grounded at the unbalanced input port 22 and the other ends of the first and third lines are the terminals 23, 24 of the balanced ports A and B. The first and the third lines 25, 26 of the second section are grounded at one end 27, while the second end of these lines are the terminals 28, 29 of the balanced ports A and B. The central line 15 is connected at one end to the unbalanced port 22 and is left unterminated at the other end. The potentials at the four balanced port terminals of this dual balun are suitable for driving elements connected to the four diodes of a star mixer, for example. It consists of purely planar coupled structures and can be realised in single-layer printed-circuit technology. It has the drawback that it is a half-wavelength long in total and therefore takes up a lot of space on a microwave monolithic integrated circuit (MMIC) of which it

is intended to form a part.

#### Summary of the Invention

In accordance with a first aspect of the invention, there is provided a balun as recited in Claim 1. Under second and third aspects of the invention, the balun according to the invention is employed respectively as part of a single-sideband mixer arrangement, as recited in Claim 5, and as part of a downconverter arrangement, as recited in Claim 11. Specific embodiments of the balun, mixer and downconverter are covered by the subclaims.

#### Brief Description of the Drawings

An embodiment of the first aspect of the invention will now be described, by way of example only, with reference to the drawings, of which:

Figure 1 is a schematic diagram of a known single balun;

Figure 2 is a schematic diagram of a known dual balun;

Figure 3 is a schematic diagram of a dual balun in accordance with the present invention;

Figure 4 is cross-sectional diagram showing the construction of an airbridge interconnect;

Figure 5 is a plan view of a microstrip layout incorporating a bypass interconnect;

Figure 6 is a circuit diagram of a monolithic single-sideband mixer incorporating a dual balun according to the invention, and

Figure 7 is a circuit diagram of a downconverter incorporating a dual balun in accordance with the invention.

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### Detailed Description of the Embodiments

Referring now to Figure 3, a dual balun according to the present invention comprises seven lines 30-36, each having a length equal to a quarter-wavelength at the mean frequency of operation of the balun. The central line 33 forms at one end the connection point for the unbalanced port 37 (Port 1) and is terminated at its other end by a connection to ground. First and second balanced ports (Ports 2 and 3, respectively) are formed from the corresponding ends of lines 30, 32 and lines 34, 36, respectively. The other end of lines 32 and 34 are grounded. Sandwiched between the lines making up each balanced port is a further line (lines 31 and 35) which is grounded at the balanced-port end. Finally, the unbalanced-port ends of lines 30 and 36 are connected by similar strip sections to the corresponding ends of lines 31 and 35, respectively, while the unbalanced-port ends of lines 31, 33 and 35 are coupled together by means of airbridges 37.

The formation of an airbridge is illustrated in Figures 4a and 4b. In Figure 4a the lines 31, 32 and 33 shown in Figure 3 are shown in cross-section on a substrate 60. To form the airbridge a photoresist 62 is deposited onto the substrate and a metallisation interconnect 61 is, in turn, deposited onto the photoresist 62 and the microstrips 31 and 33. Finally, the photoresist 62 is etched away to leave a airgap 63 (see Figure 4b). An alternative to the use of an airbridge is the use of a bypass interconnect, as illustrated in Figure 5. Here a bypass track 64 is deposited onto the substrate (not shown) carrying the strips 30-36 using the thin-film manufacturing technique.

The embodiment illustrated in Figure 3 shows symmetry of line width (parameter  $w$ ) and line spacing (parameter  $s$ ) between the two halves of the balun centred on line

33. This measure provides equal power distribution between the balanced ports. If unequal distribution were to be required in a particular application, then these parameters may be made to differ between the two halves. In practice, the actual values of line width and line spacing will be determined by the need to match the balun to the driving and driven circuitry to which it is connected, in accordance with principles well known to those skilled in the art. In practice, these dimensions may well be determined using spectral-domain techniques described in the paper "Spectral Domain Emittance Approach for Dispersion Characteristics of Generalised Printed Transmission Lines" by T. Itoh, IEEE Transactions, MTT-287, July 1980, pp 733-738. As regards the coupling of the balun to external circuitry, it may be employed as a matching network interfacing with other circuitry, as well as a balun as such. In this case it will also act as an impedance transformer.

An application of the balun according to the present invention is depicted in Figure 6. In Figure 6 block 40 represents the dual balun described above having lines 30-36 configured to form Ports 1, 2 and 3 as shown in Figure 3. The balanced ports, Ports 2 and 3, each feed an arrangement of two diodes connected in series. These are diodes 41 and 42 for Port 2 and diodes 43 and 44 for Port 3. The junctions 45, 46 of the two diode arrangements are taken to the inputs of respective low-pass filters 47, 48 and to the inputs of respective high-pass filters 49, 50. The outputs of the low-pass filters form the IF input ports 51, 52 of the mixer. Thus these filters isolate the ports 51, 52 from high frequencies within the mixer. Input 51 is an in-phase signal, while input 52 is in quadrature thereto. There therefore appear at the inputs of the high-pass filters 49, 50 a pair of sidebands of frequency  $f_{LO} + f_{IF}$ ,  $f_{LO} - f_{IF}$  and  $f_{LO} + f_{IF} + 90^\circ$ ,  $f_{LO} + f_{IF} + 90^\circ$ ,

respectively. The outputs of the high-pass filters (which may conveniently be constituted by simple capacitors) are fed into a 90-degree hybrid coupler, which in the illustrated example is a Lange coupler 53. The Lange coupler is used to separate the upper sideband 54 from the lower sideband 55. The sideband outputs are unbalanced, like the local-oscillator input port, Port 1.

The low-pass filter should preferably have at least 20dB attenuation for the local-oscillator and RF signals and less than 0.3dB attenuation for the IF signals. Conversely, the high-pass filter should preferably have 20dB attenuation for the IF signals and less than 0.3dB attenuation for the local-oscillator and RF signals.

Sufficient isolation between the local-oscillator and RF-input sections of the mixer is ensured by arranging for the diode-pairs to be excited in their odd mode. (This makes Port 1 appear as ground as far as the RF inputs are concerned and makes the diode junctions 45, 46 appear as ground as far as the local oscillator input is concerned). As well as feeding the RF input signals (which may be at IF frequency in practice) into the mixer, the low-pass filters 47 and 48 may also pass the DC voltage levels necessary to bias the diodes. The filters 47, 48 may take the form of spiral inductors or multiple-pole filters, depending on the bandwidth of the IF signals.

In addition to the illustrated mixer arrangement, the same circuit can be employed as a down converter. A suitable circuit arrangement is shown in Figure 7, in which the mixer circuit shown in Figure 6 is represented by a functional block 70. This time, however, what were the IF inputs of the mixer are now outputs feeding an additional 90° IF-hybrid 71. Likewise, the outputs 54, 55 of the 90° hybrid (e.g. Lange coupler) in the former mixer are now inputs taken to an RF input signal and a ground-

referenced load 72, respectively. Finally, one of the outputs of the  $90^\circ$  hybrid 71 forms an IF output for the downconverter arrangement, while the other output is terminated in a matching load 73. (The image frequency of the RF input appears at this output and is absorbed in the load). The local oscillator input (LO) of the former mixer (block 70) continues to be the local oscillator input for the downconverter.

The balun, mixer and downconverter described can be implemented in MIC (microwave integrated circuit) or MMIC (monolithic microwave integrated circuit) technology.

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